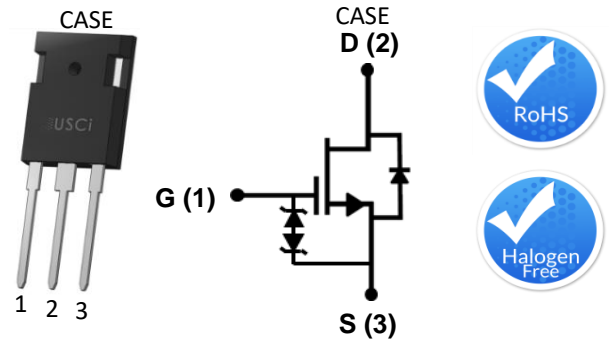


## Description

United Silicon Carbide's cascode products co-package its high-performance G3 SiC JFETs with a cascode optimized MOSFET to produce the only standard gate drive SiC device in the market today. This series exhibits ultra-low gate charge, but also the best reverse recovery characteristics of any device of similar ratings. These devices are excellent for switching inductive loads when used with recommended RC-snubbers, and any application requiring standard gate drive.



Part Number	Package	Marking
UF3C065040K3S	TO-247-3L	UF3C065040K3S

## Features

- ◆ Typical on-resistance  $R_{DS(on),typ}$  of 42mΩ
- ◆ Maximum operating temperature of 175°C
- ◆ Excellent reverse recovery
- ◆ Low gate charge
- ◆ Low intrinsic capacitance
- ◆ ESD protected, HBM class 2
- ◆ Very low switching losses (required RC-snubber loss negligible under typical operating conditions)

## Typical Applications

- ◆ EV charging
- ◆ PV inverters
- ◆ Switch mode power supplies
- ◆ Power factor correction modules
- ◆ Motor drives
- ◆ Induction heating

## Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	$V_{DS}$		650	V
Gate-source voltage	$V_{GS}$	DC	-25 to +25	V
Continuous drain current <sup>1</sup>	$I_D$	$T_C=25^\circ\text{C}$	54	A
		$T_C=100^\circ\text{C}$	40	A
Pulsed drain current <sup>2</sup>	$I_{DM}$	$T_C=25^\circ\text{C}$	125	A
Single pulsed avalanche energy <sup>3</sup>	$E_{AS}$	$L=15\text{mH}$ , $I_{AS}=3.19\text{A}$	76	mJ
Power dissipation	$P_{tot}$	$T_C=25^\circ\text{C}$	326	W
Maximum junction temperature	$T_{J,max}$		175	°C
Operating and storage temperature	$T_J$ , $T_{STG}$		-55 to 175	°C
Max. lead temperature for soldering, 1/8" from case for 5 seconds	$T_L$		250	°C

1 Limited by  $T_{J,max}$

2 Pulse width  $t_p$  limited by  $T_{J,max}$

3 Starting  $T_J = 25^\circ\text{C}$

**Electrical Characteristics** ( $T_J = +25^\circ\text{C}$  unless otherwise specified)

**Typical Performance - Static**

Parameter	Symbol	Test Conditions	Value			Units
			Min	Typ	Max	
Drain-source breakdown voltage	$BV_{DS}$	$V_{GS}=0V, I_D=1mA$	650			V
Total drain leakage current	$I_{DSS}$	$V_{DS}=650V,$ $V_{GS}=0V, T_J=25^\circ\text{C}$		0.7	150	$\mu\text{A}$
		$V_{DS}=650V,$ $V_{GS}=0V, T_J=175^\circ\text{C}$		10		
Total gate leakage current	$I_{GSS}$	$V_{DS}=0V, T_J=25^\circ\text{C},$ $V_{GS}=-20V / +20V$		6	$\pm 20$	$\mu\text{A}$
Drain-source on-resistance	$R_{DS(on)}$	$V_{GS}=12V, I_D=40A,$ $T_J=25^\circ\text{C}$		42	52	$\text{m}\Omega$
		$V_{GS}=12V, I_D=40A,$ $T_J=175^\circ\text{C}$		78		
Gate threshold voltage	$V_{G(th)}$	$V_{DS}=5V, I_D=10mA$	4	5	6	V
Gate resistance	$R_G$	$f=1\text{MHz}, \text{open drain}$		4.5		$\Omega$

**Typical Performance - Reverse Diode**

Parameter	Symbol	Test Conditions	Value			Units
			Min	Typ	Max	
Diode continuous forward current <sup>1</sup>	$I_S$	$T_C=25^\circ\text{C}$			54	A
Diode pulse current <sup>2</sup>	$I_{S,pulse}$	$T_C=25^\circ\text{C}$			125	A
Forward voltage	$V_{FSD}$	$V_{GS}=0V, I_F=20A,$ $T_J=25^\circ\text{C}$		1.5	1.75	V
		$V_{GS}=0V, I_F=20A,$ $T_J=175^\circ\text{C}$		1.8		
Reverse recovery charge	$Q_{rr}$	$V_R=400V, I_F=40A,$ $V_{GS}=-5V, R_{G\_EXT}=20\Omega$		138		nC
Reverse recovery time	$t_{rr}$	$di/dt=1100A/\mu\text{s},$ $T_J=25^\circ\text{C}$		38		ns
Reverse recovery charge	$Q_{rr}$	$V_R=400V, I_F=40A,$ $V_{GS}=-5V, R_{G\_EXT}=20\Omega$		137		nC
Reverse recovery time	$t_{rr}$	$di/dt=1100A/\mu\text{s},$ $T_J=150^\circ\text{C}$		38		ns

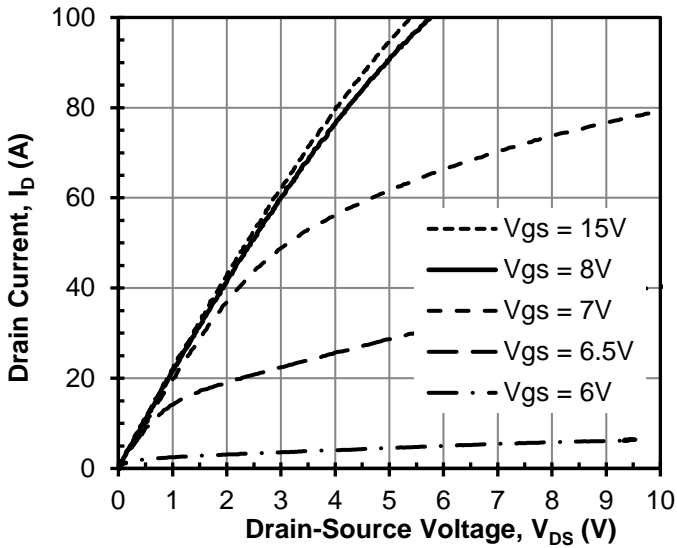
**Typical Performance - Dynamic**

Parameter	symbol	Test Conditions	Value			Units		
			Min	Typ	Max			
Input capacitance	$C_{iss}$	$V_{DS}=100V,$ $V_{GS}=0V,$ $f=100kHz$		1500		pF		
Output capacitance	$C_{oss}$			200				
Reverse transfer capacitance	$C_{rss}$			2.2				
Effective output capacitance, energy related	$C_{oss(er)}$	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		146		pF		
Effective output capacitance, time related	$C_{oss(tr)}$	$V_{DS}=0V$ to 400V, $V_{GS}=0V$		325		pF		
$C_{oss}$ stored energy	$E_{oss}$	$V_{DS}=400V,$ $V_{GS}=0V$		11.7		μJ		
Total gate charge	$Q_G$	$V_{DS}=400V,$ $I_D=40A,$ $V_{GS}=-5V$ to 15V		51		nC		
Gate-drain charge	$Q_{GD}$			11				
Gate-source charge	$Q_{GS}$			19				
Turn-on delay time	$t_{d(on)}$	$V_{DS}=400V,$ $I_D=40A,$ Gate Driver=-5V to +15V, Turn-on $R_{G,EXT}=1.8\Omega,$ Turn-off $R_{G,EXT}=22\Omega$ Inductive Load, FWD: same device with $V_{GS} = -5V$ and $R_G = 22\Omega$ RC snubber: $R_S=5\Omega$ and $C_S=150pF$ $T_J=25^\circ C$		35		ns		
Rise time	$t_r$			24				
Turn-off delay time	$t_{d(off)}$			57				
Fall time	$t_f$			14				
Turn-on energy including $R_S$ energy <sup>4</sup>	$E_{ON}$	$V_{DS}=400V,$ $I_D=40A,$ Gate Driver=-5V to +15V, Turn-on $R_{G,EXT}=1.8\Omega,$ Turn-off $R_{G,EXT}=22\Omega$ Inductive Load, FWD: same device with $V_{GS} = -5V$ and $R_G = 22\Omega$ RC snubber: $R_S=5\Omega$ and $C_S=150pF$ $T_J=25^\circ C$		500		μJ		
Turn-off energy including $R_S$ energy <sup>4</sup>	$E_{OFF}$			118				
Total switching energy including $R_S$ energy <sup>4</sup>	$E_{TOTAL}$			618				
Snubber $R_S$ energy during turn-on	$E_{RS\_ON}$			1.7				
Snubber $R_S$ energy during turn-off	$E_{RS\_OFF}$			4.5				
Turn-on delay time	$t_{d(on)}$		$V_{DS}=400V,$ $I_D=40A,$ Gate Driver=-5V to +15V, Turn-on $R_{G,EXT}=1.8\Omega,$ Turn-off $R_{G,EXT}=22\Omega$ Inductive Load, FWD: same device with $V_{GS} = -5V$ and $R_G = 22\Omega$ RC snubber: $R_S=5\Omega$ and $C_S=150pF$ $T_J=150^\circ C$		35			ns
Rise time	$t_r$				22			
Turn-off delay time	$t_{d(off)}$				60			
Fall time	$t_f$			13				
Turn-on energy including $R_S$ energy <sup>4</sup>	$E_{ON}$			479				
Turn-off energy including $R_S$ energy <sup>4</sup>	$E_{OFF}$			124				
Total switching energy including $R_S$ energy <sup>4</sup>	$E_{TOTAL}$			603				
Snubber $R_S$ energy during turn-on	$E_{RS\_ON}$			1.8				
Snubber $R_S$ energy during turn-off	$E_{RS\_OFF}$		5.3					

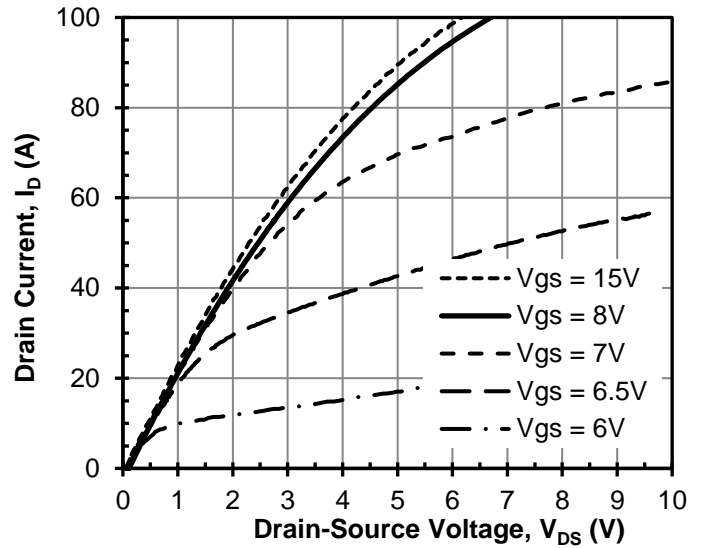
4 The switching performance are evaluated with a RC snubber circuit as shown in Figure 24.

**Thermal Characteristics**

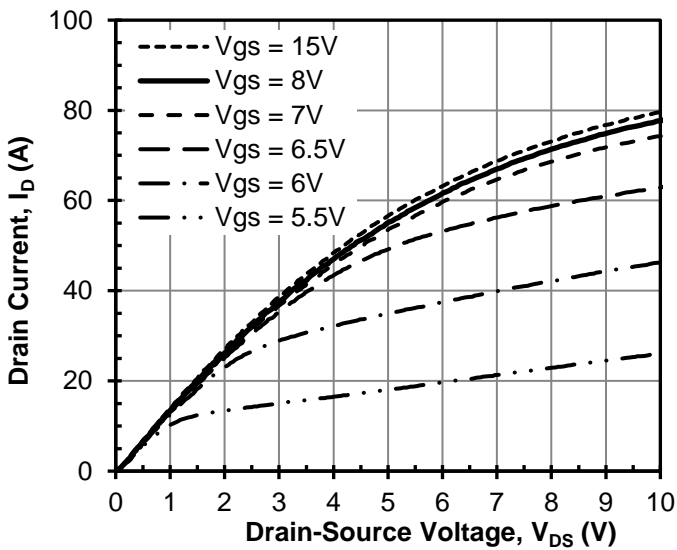
Parameter	symbol	Test Conditions	Value			Units
			Min	Typ	Max	
Thermal resistance, junction-to-case	$R_{\theta JC}$			0.35	0.46	°C/W

**Typical Performance Diagrams**


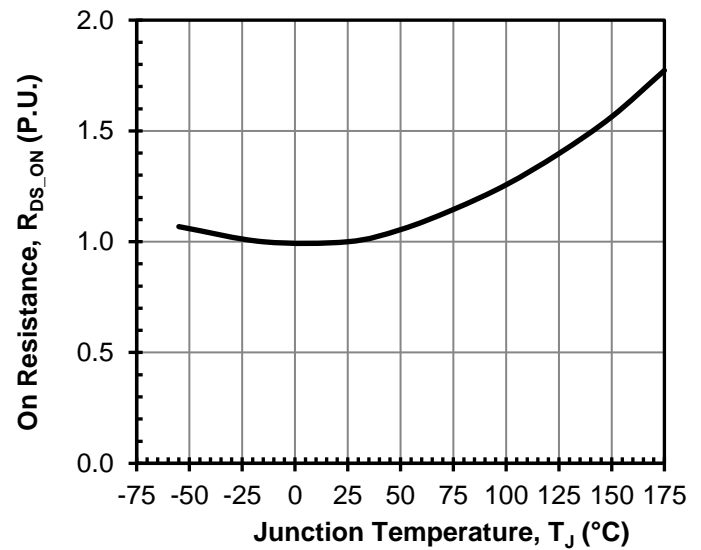
**Figure 1** Typical output characteristics  
at  $T_J = -55^\circ\text{C}$ ,  $t_p < 250 \mu\text{s}$



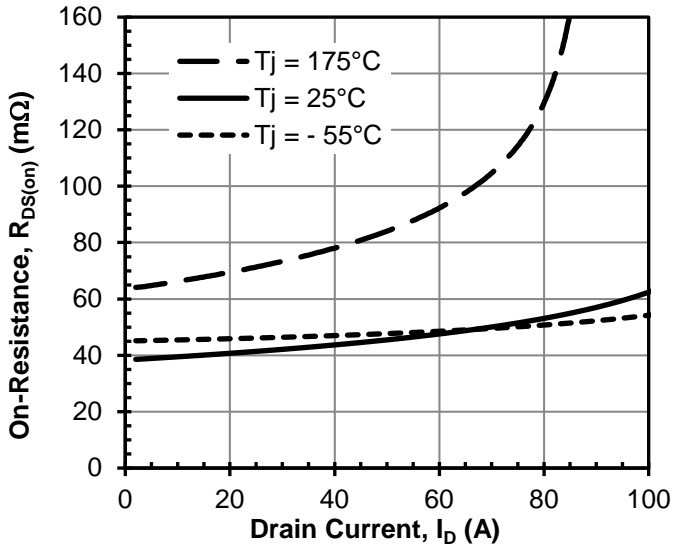
**Figure 2** Typical output characteristics  
at  $T_J = 25^\circ\text{C}$ ,  $t_p < 250 \mu\text{s}$



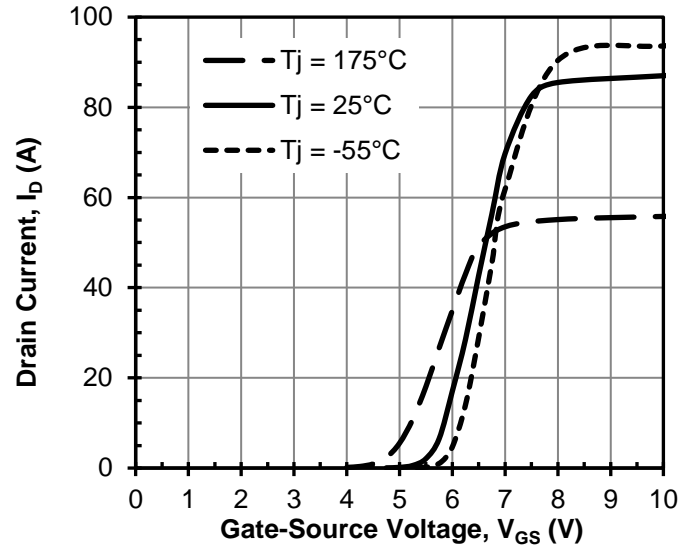
**Figure 3** Typical output characteristics  
at  $T_J = 175^\circ\text{C}$ ,  $t_p < 250 \mu\text{s}$



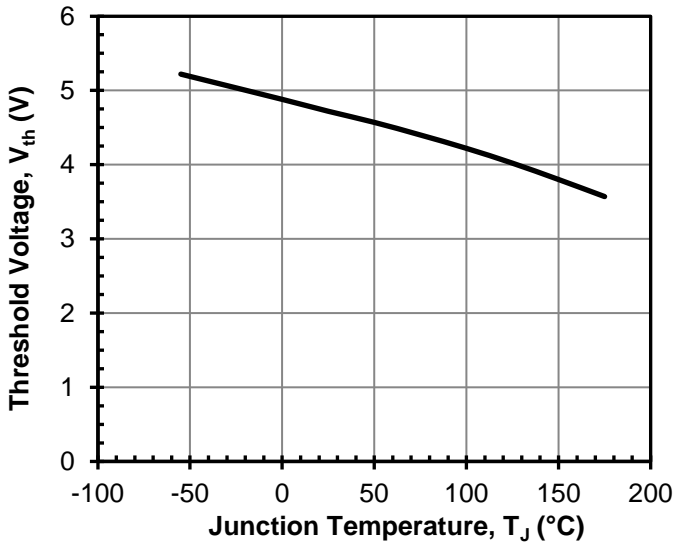
**Figure 4** Normalized on-resistance vs.  
temperature at  $V_{GS} = 12\text{V}$  and  
 $I_D = 40\text{A}$



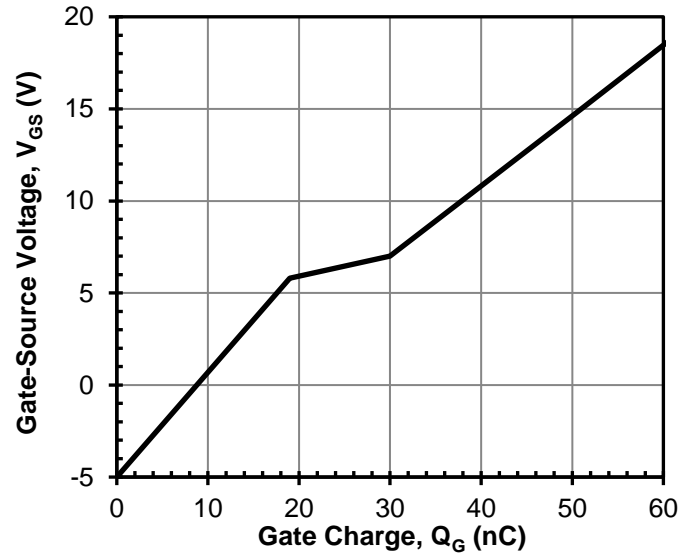
**Figure 5** Typical drain-source on-resistance at  $V_{GS} = 12V$



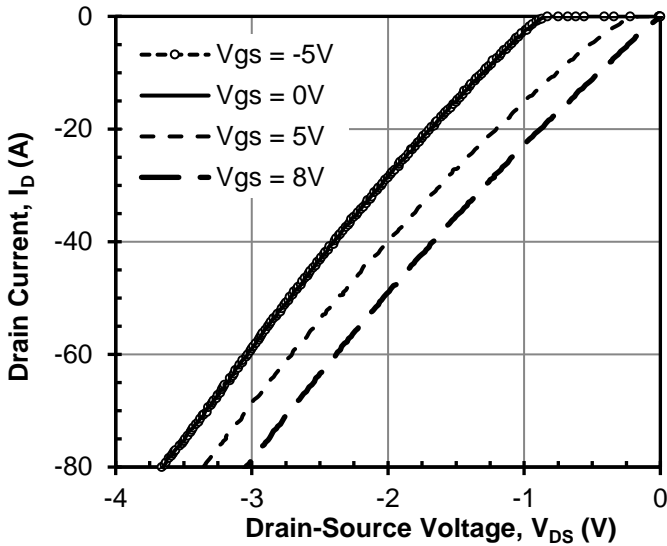
**Figure 6** Typical transfer characteristics at  $V_{DS} = 5V$



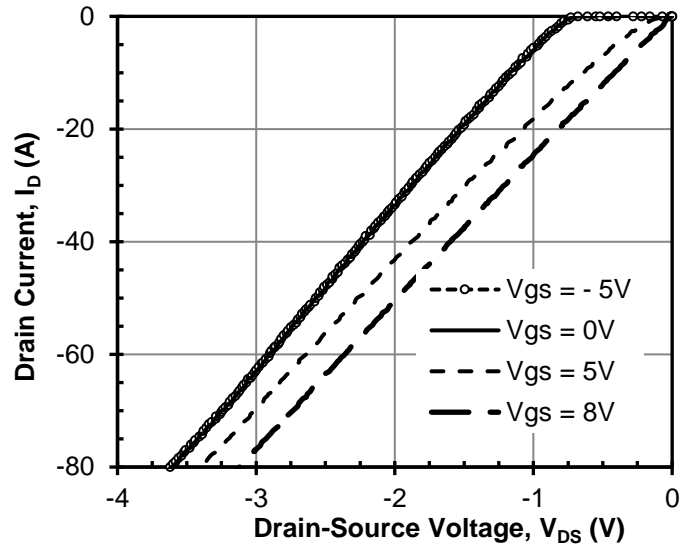
**Figure 7** Threshold voltage vs.  $T_j$  at  $V_{DS} = 5V$  and  $I_D = 10mA$



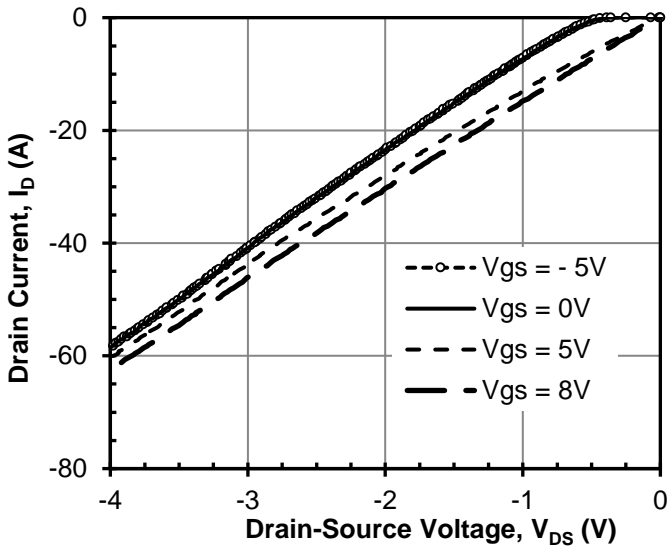
**Figure 8** Typical gate charge at  $V_{DS} = 400V$  and  $I_D = 40A$



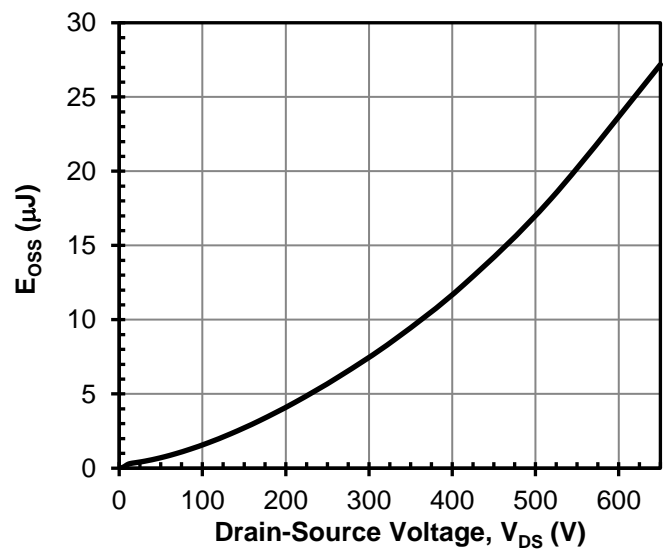
**Figure 9** 3rd quadrant characteristics  
at  $T_J = -55^\circ\text{C}$



**Figure 10** 3rd quadrant characteristics  
at  $T_J = 25^\circ\text{C}$



**Figure 11** 3rd quadrant characteristics  
at  $T_J = 175^\circ\text{C}$



**Figure 12** Typical stored energy in  $C_{OSS}$   
at  $V_{GS} = 0\text{V}$

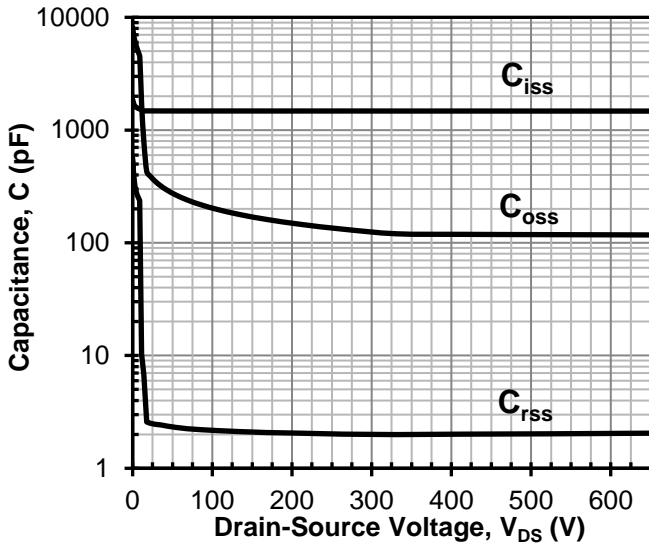


Figure 13 Typical capacitances at 100kHz and  $V_{GS} = 0V$

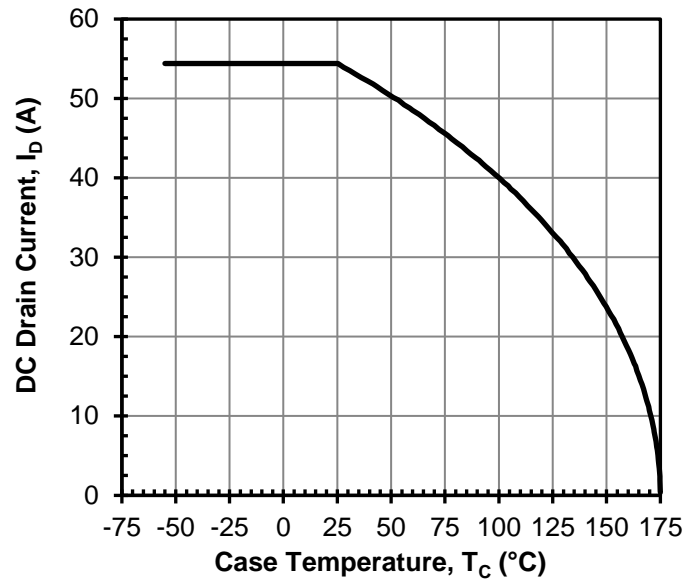


Figure 14 DC drain current derating

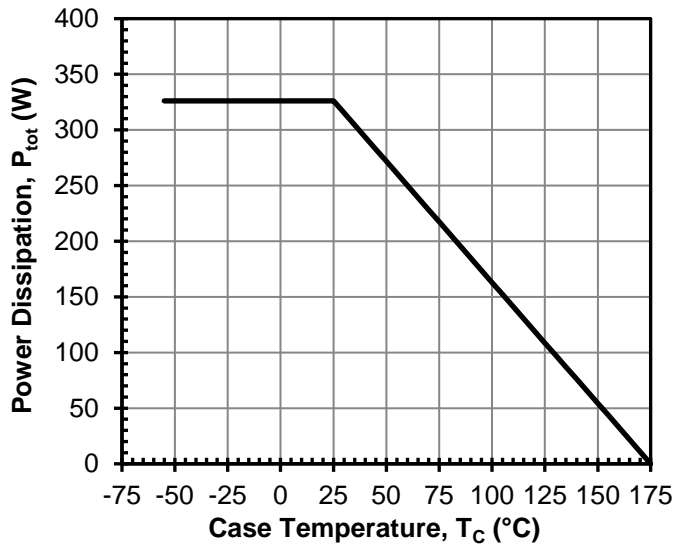


Figure 15 Total power dissipation

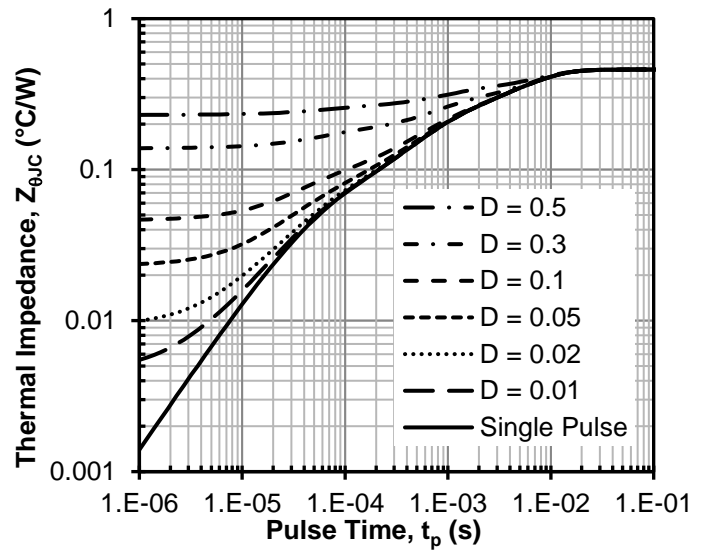
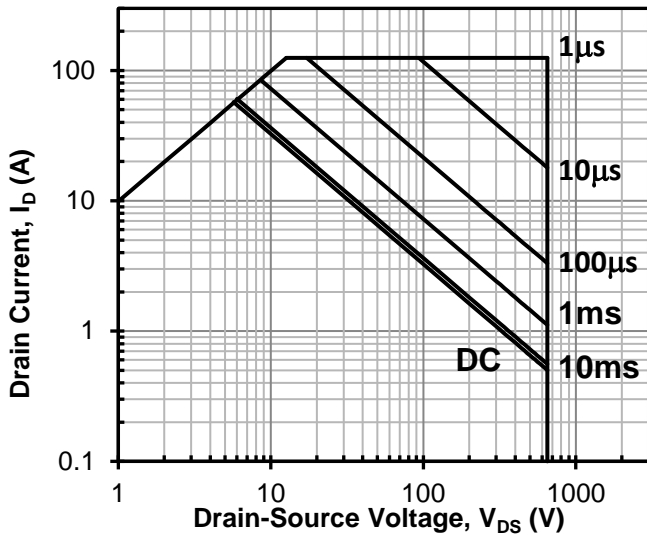
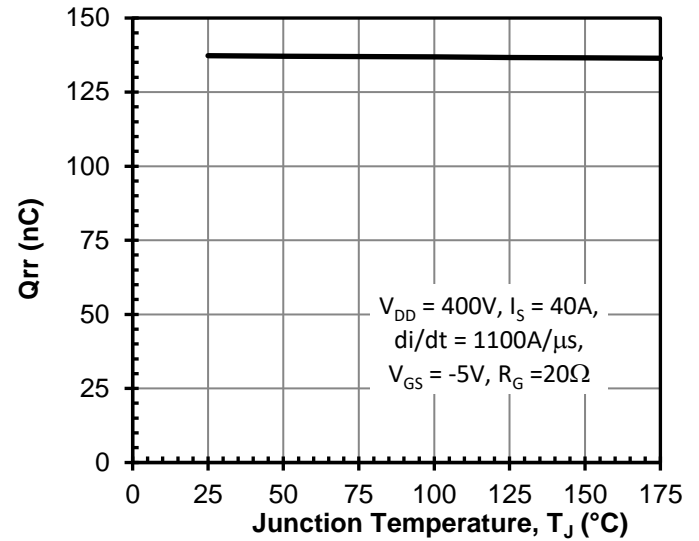


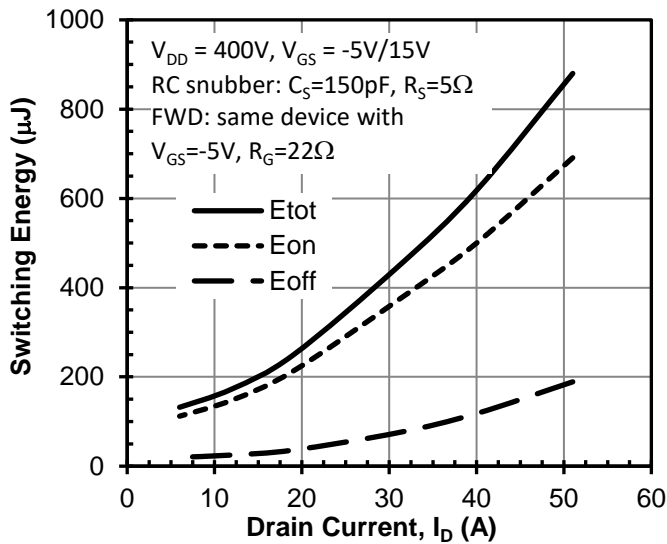
Figure 16 Maximum transient thermal impedance



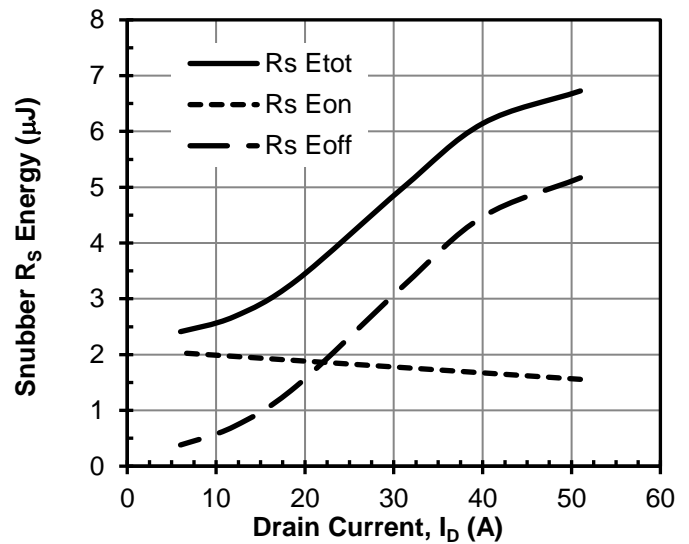
**Figure 17 Safe operation area**  
 $T_c = 25^\circ\text{C}$ ,  $D = 0$ , Parameter  $t_p$



**Figure 18 Reverse recovery charge  $Q_{rr}$  vs. junction temperature**



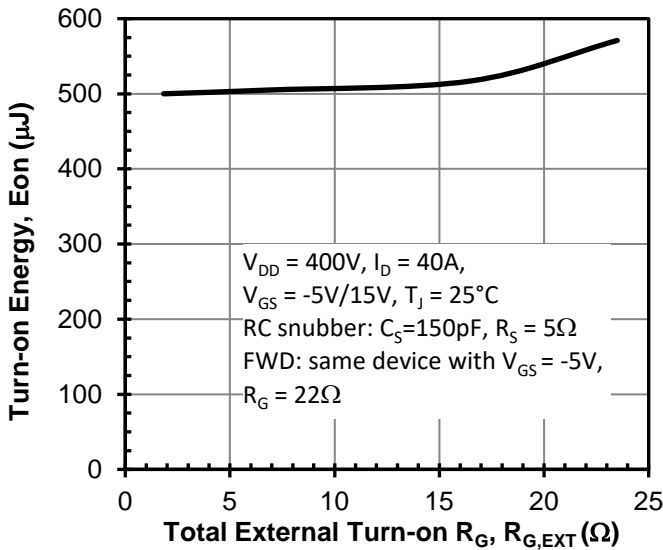
(a)



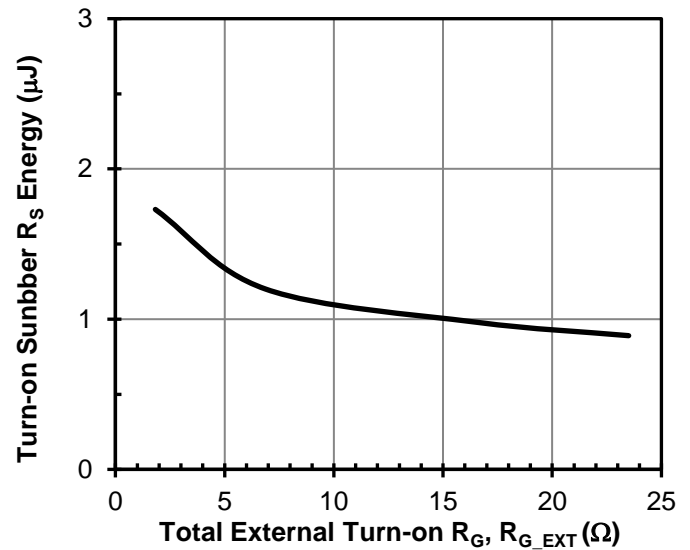
(b)

**Figure 19 Clamped inductive switching energy (a) and RC snubber energy loss (b) vs. drain current at  $T_J = 25^\circ\text{C}$ , turn-on  $R_{G\_EXT} = 1.8\Omega$  and turn-off  $R_{G\_EXT} = 22\Omega$**



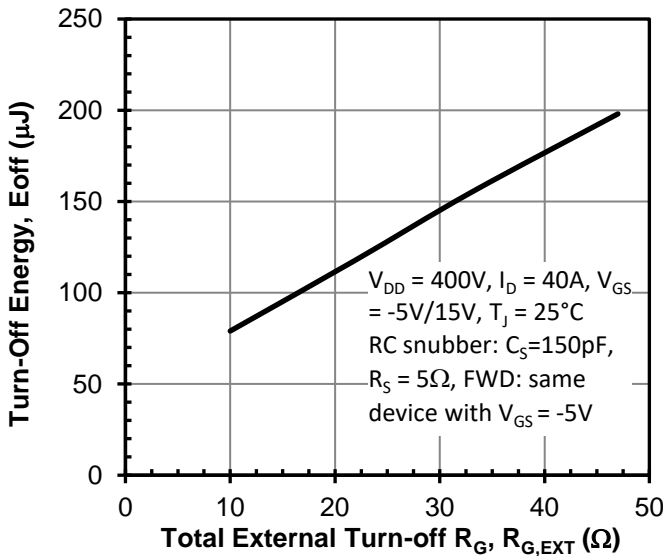


(a)

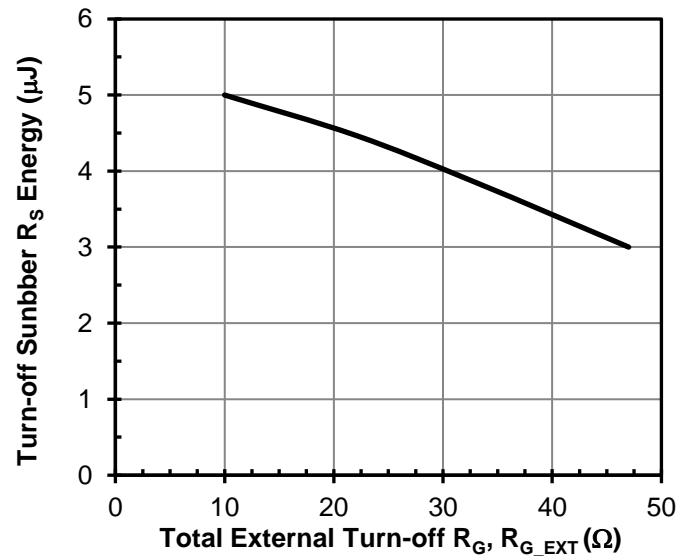


(b)

**Figure 20** Clamped inductive switching turn-on energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of total external turn-off gate resistor  $R_{G,EXT}$ .

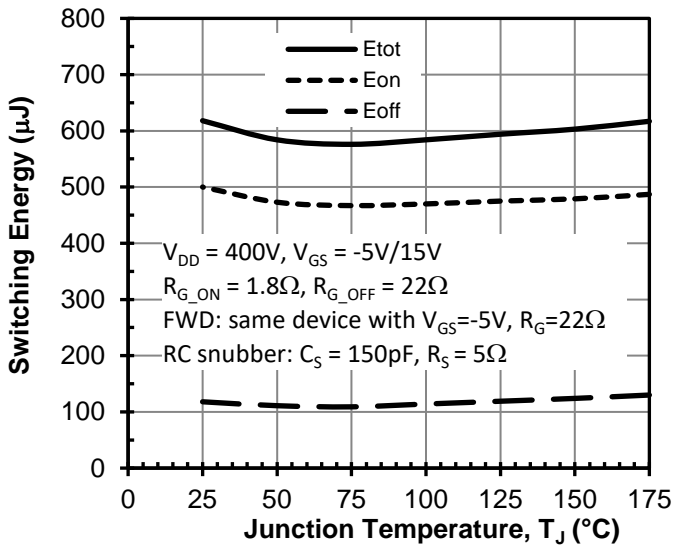


(a)

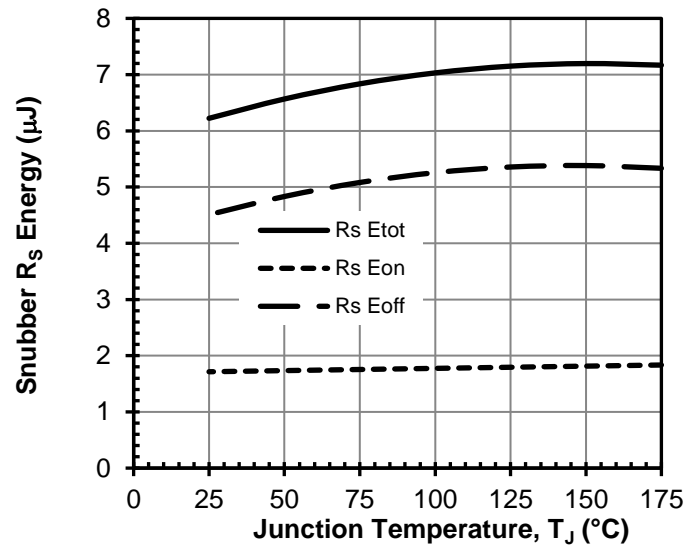


(b)

**Figure 21** Clamped inductive switching turn-off energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of total external turn-off gate resistor  $R_{G,EXT}$ .

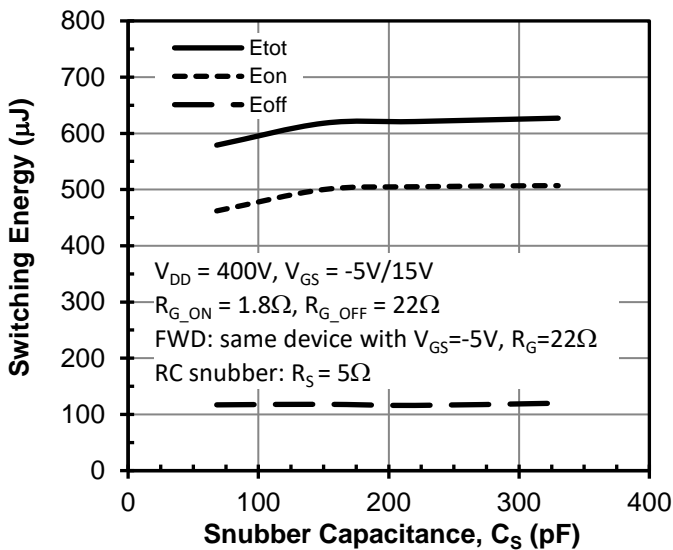


(a)

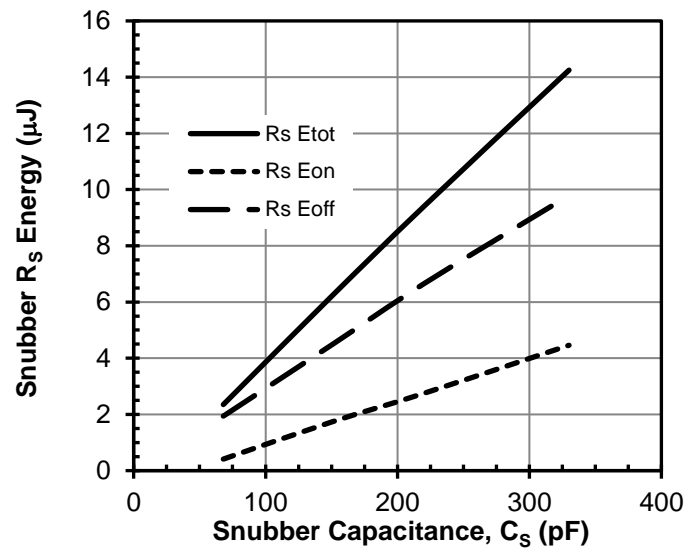


(b)

**Figure 22** Clamped inductive switching energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of junction temperature at  $I_D = 40A$

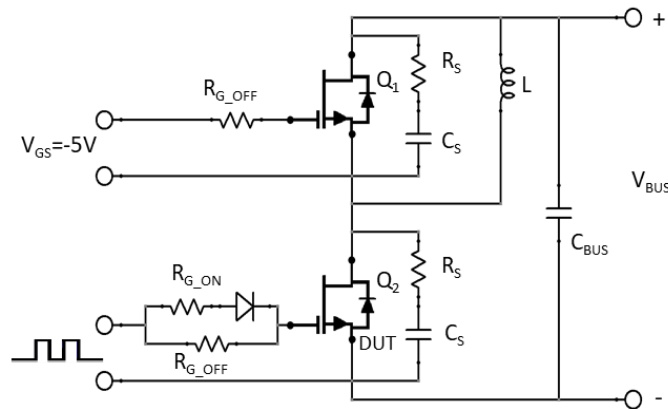


(a)



(b)

**Figure 23** Clamped inductive switching energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of the snubber capacitance  $C_s$  at  $I_D = 40A$  and  $T_J = 25^\circ C$



**Figure 24 Inductive load switching test circuit**

**An RC snubber ( $R_S = 5\Omega$ ,  $C_S = 150\text{pF}$ ) is required to improve the turn-off waveforms.**

## Applications Information

SiC cascodes are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ( $R_{DS(on)}$ ), output capacitance ( $C_{oss}$ ), gate charge ( $Q_g$ ), and reverse recovery charge ( $Q_{rr}$ ) leading to low conduction and switching losses. The SiC cascodes also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high  $dv/dt$  and  $di/dt$  rates. An external gate resistor is recommended when the cascode is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on cascode operation, see [www.unitedsic.com](http://www.unitedsic.com).

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